

**SYSTEM AND METHOD OF COMMUNICATION USING
TRANSMIT ANTENNA DIVERSITY BASED UPON
UPLINK MEASUREMENTS FOR THE TDD MODE OF WCDMA**

5 Claim To Priority Of Provisional Application

 The application claims priority under 35 U.S.C. § 119(e)(1) of provisional application serial number 60/146,619, filed 07/30/99, by Timothy M. Schmidl and Anand G. Dabak and further claims priority under 35 U.S.C. §119(e)(1) of provisional application serial number 60/146,477, filed 07/30/99, by Timothy M. Schmidl and Anand
10 G. Dabak.

Background of the Invention

1. Field of the Invention

15 This invention relates generally to wideband code division multiple access (WCDMA) for a communication system and more particularly to a system and method of data communication using transmit antenna diversity based on uplink measurements for the time division duplex (TDD) mode of WCDMA.

20 2. Description of the Prior Art

 Present code division multiple access (CDMA) systems are characterized by simultaneous transmission of different data signals over a common channel by assigning each signal a unique code. This unique code is matched with a code of a selected receiver, i.e., mobile terminal, to determine the proper recipient of a data signal. These
25 different data signals arrive at the receiver via multiple paths due to ground clutter and unpredictable signal reflection. Additive effects of these multiple data signals at the receiver may result in significant fading or variation in received signal strength. In general, this fading due to multiple data paths may be diminished by spreading the transmitted energy over a wide bandwidth. This wide bandwidth results in greatly
30 reduced fading compared to narrow band transmission modes such as frequency division multiple access (FDMA) or time division multiple access (TDMA).

New standards are continually emerging for next generation wideband code division multiple access (WCDMA) communication systems as described in U.S. Patent Application No. 09/205,029, filed December 3, 1998, and incorporated herein by reference. Therein, Dabak et al. describe a method of space time transmit diversity (STTD) for frequency division duplex (FDD) WCDMA systems. These FDD systems are coherent communications systems with pilot symbol assisted channel estimation schemes. These pilot symbols are transmitted as quadrature phase shift keyed (QPSK) known data in predetermined time frames to any receivers within range. The frames may propagate in discontinuous transmission (DTX) mode. For voice traffic, transmission of user data occurs when the user speaks, but no data symbol transmission occurs when the user is silent. Similarly for packet data, the user data may be transmitted only when packets are ready to be sent. The frames include pilot symbols as well as other control symbols such as transmit power control (TPC) symbols and rate information (RI) symbols. These control symbols include multiple bits otherwise known as chips to distinguish them from data bits. The chip transmission time (T_c), therefore, is equal to the symbol rate (T) divided by the number of chips in the symbol (G).

Time division duplex (TDD) provides an alternative communication standard for WCDMA, FDD systems. TDD data are transmitted as QPSK symbols in data packets of a predetermined duration or time slot. Each data packet includes a training sequence or midamble within the time slot. Data packets are exchanged within a cell formed by a base station in communication with nearby mobile units. Data in adjacent cells are modulated by different periodic codes. The midamble is formed by adding time shifted versions of the same basic sequence with each time shift corresponding to a mobile unit within the cell. The spreading factor (SF) or chips per symbol of the modulation is preferably sixteen or less. Since the periodic code within the cell is the same and the spreading factor is small, however, interference from the base station and other mobile units within the cell is not received as Gaussian noise. Typical matched filter circuits used in FDD systems, therefore, are unsuitable for eliminating this intra cell interference. A solution to this problem was presented by Anja Klein et al., *Zero Forcing and Minimum Mean-Square-Error Equalization for Multiuser Detection in Code-Division Multiple-Access Channels*, IEEE Transactions on Vehicular Technology, 276-287 (1996).

Therein, Klein et al. teach zero forcing (ZF) and minimum mean-square-error (MMSE) equalization with and without decision feedback (DF) to reduce both inter-symbol interference (ISI) and multiple-access interference (MAI). Klein et al. further cites P. Jung, J. Blanz and P.W. Baier, *Coherent Receiver Antenna Diversity for CDMA Mobile Radio Systems Using Joint Detection*, Proc. IEEE Int. Symp. Pers. Indoor and Mobile Radio Communications, 488-492 (1993), for the proposition that these techniques may be used in combination with antenna diversity.

Delay diversity is a method of transmit antenna diversity in which the same signal is transmitted from multiple antennas, with each antenna having a different time delay.

Delay diversity was introduced for TDMA systems such as IS-54 and GSM (global system for mobile communications) by N. Seshadri and J. Winters, *Two Signaling Schemes for Improving the Error Performance of Frequency-Division-Duplex (FDD) Transmission Systems Using Transmitter Diversity*, Vehicular Technology Conference, pp. 508-511 (1993) and A. Wittneben, *A New Bandwidth Efficient Transmit Antenna Modulation Diversity Scheme for Linear Digital Modulation*, International Communications Conference, vol. 3, pp. 1630-1634 (1993), wherein the delays for each antenna are chosen to be a multiple of the symbol interval. Delay diversity for a synchronous CDMA system (IS-95) was disclosed in U.S. Patent No. 5,781,541, entitled *CDMA System Having Time-Distributed Transmission Paths for Multipath Reception*, by A. Schneider, issued July 14, 1998 wherein the delays for each antenna are greater than a chip interval and less than the base station sequence offset between base stations. The Seshadri et al., Wittneben and Schneider references are incorporated by reference herein. Delay diversity for another CDMA communication system and using a distributed antenna system to provide multipath signals in order to facilitate signal diversity for enhanced system performance was disclosed by Gilhousen, et al. in U.S. Patent No. 5,280,472, entitled *CDMA Microcellular Telephone System and Distributed Antenna System Therefor*, issued January 18, 1994. A communication system having the advantages of delay diversity associated with TDD base terminals in combination with joint detection at the mobile terminals is, however, presently unknown. In view of the above, a wireless communication system having variable transmit antenna delays based on uplink measurements associated with the TDD mode of WCDMA in combination with

joint detection of the transmitted signals at the mobile terminal is both advantageous and desirable.

Summary of the Invention

The present invention is directed to a communication system using variable transmit antenna delays associated with multiple antennas, with each antenna having a different time delay, based upon uplink measurements for the TDD mode of WCDMA. The signals for each channel are delayed at baseband, so different delays can be used for each channel. Since the base station must measure the delay profile in the uplink to receive the uplink transmission, there is no additional complexity involved in measuring the positions of the paths in the uplink for the TDD mode. Since this system is a time division duplex (TDD) system, the uplink and downlink profiles are the same. The delay between the antennas can be chosen so that the strongest paths do not overlap so that full diversity can be achieved. For broadcast channels transmitted to all mobile terminals, there is less flexibility in choosing the delays between antennas since the same delay is used for all users. In this case, the antenna delays can be chosen so as to benefit the maximum number of users or the users at the edge of the cell.

Thus, a method of communicating data between a base station having a plurality of antennas and at least one mobile terminal according to one embodiment of the present invention comprises the steps of:

transmitting from the base station, derived versions of a midamble signal to each antenna within the plurality of antennas; and
providing a distinct delay associated with each derived version of the midamble signal and its respective antenna.

Another embodiment of the present method for communicating data between a base station having a plurality of antennas and at least one mobile terminal comprises the steps of:

receiving at the base station, a time division duplex mode uplink signal from each mobile terminal in communication with the base station and estimating a path profile associated with each received uplink signal;
transmitting from the base station, a time division duplex signal to each antenna within the plurality of antennas; and

providing a distinct delay associated with each time division duplex signal and its respective antenna.

A communication system according to one embodiment of the present inventions comprises:

- 5 a base station having a plurality of spaced apart antennas;
- signal distribution means for coupling time division duplex communication signals between the base station and the plurality of spaced apart antennas; and
- delay means operatively coupled to the antennas and the signal distribution means for providing a distinct delay in each of the time division duplex communication signals
- 10 coupled between the base station and the plurality of spaced apart antennas.

A data communication system according to another embodiment of the present invention comprises:

- a base station having a plurality of spaced apart antennas;
- at least one mobile terminal in communication with the base station; and
- 15 means for providing a distinct delay associated with each antenna such that a time division duplex communication signal coupled between the base station and the plurality of spaced apart antennas can be demodulated within the at least one mobile terminal.

A data communication system according to yet another embodiment of the present invention comprises:

- 20 a base station having a plurality of spaced apart antennas;
- means for transmitting from the base station, derived versions of a midamble signal to each antenna within the plurality of spaced apart antennas; and
- means for providing a distinct delay associated with each derived version of the midamble signal and its respective antenna within the plurality of spaced apart antennas.

25 A data communication system according to still another embodiment of the present invention comprises:

- a base station configured to communicate with the at least one mobile terminal, the base station having a plurality of spaced apart antennas and further having:
- a data processor;
- 30 a data input device in communication with the data processor;
- an algorithmic software directing the data processor; and

a data storage unit, wherein discrete signal uplink data associated with at least one mobile terminal in communication with the base station is stored and supplied to the data processor such that the data processor, directed by the algorithmic software, can automatically derive midamble signal parameters using algorithmically defined relationships associated with the discrete signal uplink data such that derived midamble signals communicated between the base unit and each respective antenna will be characterized by at least one distinct signal parameter selected from the group consisting of signal delay, signal phase and signal amplitude.

A time-division duplex (TDD) data communication system according to yet another embodiment of the present invention in which system users communicate information signals through a base station using TDD communication signals, comprises a base station having an antenna system comprising:

a plurality of spaced apart antennas;

signal distribution means for coupling TDD communication signals between a base station and the plurality of spaced apart antennas; and

variable delay means operatively coupled to the plurality of spaced apart antennas and the signal distribution means for providing derived delays associated with the TDD communication signals and the plurality of spaced apart antennas.

As used herein, the following words have the following meanings. The words “algorithmic software” means an algorithmic program used to direct the processing of data by a computer or data processing device. The words “data processing device” as used herein refer to a CPU, DSP, microprocessor, micro-controller, or other like device and an interface system. The interface system provides access to the data processing device such that data could be entered and processed by the data processing device. The words “discrete data” as used herein are interchangeable with “digitized data” and “digitized data” as used herein means data which are stored in the form of singularly isolated, discontinuous data or digits.

In another aspect of the invention, a multiple transmit antenna base station utilizes delay diversity to deliver TDD signals to a mobile terminal.

In yet another aspect of the invention, a multiple transmit antenna base station utilizes delay diversity to deliver signals to a mobile terminal having a joint detector to accommodate signal demodulation associated with the TDD mode of WCDMA.

5 In another aspect of the invention, a multiple transmit antenna base station utilizes adaptive variable delay diversity to accommodate the TDD mode of WCDMA data communication.

In yet another aspect of the invention, a multiple transmit antenna base station utilizes adaptive variable delay diversity in combination with a training midamble to generate multiple communication channel delay characteristics.

10 In still another aspect of the invention, a communication system comprises a multiple transmit antenna base station utilizing adaptive variable delay diversity in combination with a joint detector at the mobile receiver to remove interference from other users on the same cell.

Brief Description of the Drawings

Other aspects and features of the present invention and many of the attendant
5 advantages of the present invention will be readily appreciated as the same become better
understood by reference to the following detailed description when considered in
connection with the accompanying drawings in which like reference numerals designate
like parts throughout the figures thereof and wherein:

Figure 1 is a block diagram of a typical cell showing a single base station in
10 communication with a plurality of mobile terminals;

Figure 2 is a diagram of a TDD radio frame;

Figure 3 is a diagram of a time slot within the radio frame of Figure 2;

Figure 4(a) is a block diagram illustrating one implementation of delay diversity
at the base station shown in Figure 1 and having K channels and N antennas suitable for
15 accommodating the TDD mode of WCDMA data communication according to one
embodiment of the present invention;

Figure 4(b) is a block diagram illustrating one implementation of delay diversity
at the base station shown in Figure 1 and having K channels and N antennas suitable for
accommodating the TDD mode of WCDMA data communication according to yet
20 another embodiment of the present invention;

Figure 5 is a block diagram illustrating a method of choosing channel delays
associated with multiple antennas according to one embodiment of the present invention;

Figure 6 is a diagram of a delay profile at a mobile terminal with a delay of 2
chips between two transmit antennas at the base station shown in Figure 1 according to
25 one embodiment of the present invention;

Figure 7 is a diagram of a delay profile at a mobile terminal with a delay of 1 chip
between two transmit antennas at the base station shown in Figure 1 according to one
embodiment of the present invention;

Figure 8 is a simulation diagram showing bit error rate (BER) performance with
30 and without delay diversity on the downlink using the Vehicular B channel model; and

Figure 9 is a simulation diagram showing bit error rate (BER) performance with and without delay diversity on the downlink using the Outdoor-to-Indoor and Pedestrian channel model.

While the above-identified drawing figures set forth alternative embodiments, other embodiments of the present invention are also contemplated, as noted in the discussion. In all cases, this disclosure presents illustrated embodiments of the present invention by way of representation and not limitation. Numerous other modifications and embodiments can be devised by those skilled in the art which fall within the scope and spirit of the principles of this invention.

10

10
11
12
13
14
15
16
17
18
19
20
21
22
23
24
25
26
27
28
29
30
31
32
33
34
35
36
37
38
39
40
41
42
43
44
45
46
47
48
49
50
51
52
53
54
55
56
57
58
59
60
61
62
63
64
65
66
67
68
69
70
71
72
73
74
75
76
77
78
79
80
81
82
83
84
85
86
87
88
89
90
91
92
93
94
95
96
97
98
99
100
101
102
103
104
105
106
107
108
109
110
111
112
113
114
115
116
117
118
119
120
121
122
123
124
125
126
127
128
129
130
131
132
133
134
135
136
137
138
139
140
141
142
143
144
145
146
147
148
149
150
151
152
153
154
155
156
157
158
159
160
161
162
163
164
165
166
167
168
169
170
171
172
173
174
175
176
177
178
179
180
181
182
183
184
185
186
187
188
189
190
191
192
193
194
195
196
197
198
199
200
201
202
203
204
205
206
207
208
209
210
211
212
213
214
215
216
217
218
219
220
221
222
223
224
225
226
227
228
229
230
231
232
233
234
235
236
237
238
239
240
241
242
243
244
245
246
247
248
249
250
251
252
253
254
255
256
257
258
259
260
261
262
263
264
265
266
267
268
269
270
271
272
273
274
275
276
277
278
279
280
281
282
283
284
285
286
287
288
289
290
291
292
293
294
295
296
297
298
299
300
301
302
303
304
305
306
307
308
309
310
311
312
313
314
315
316
317
318
319
320
321
322
323
324
325
326
327
328
329
330
331
332
333
334
335
336
337
338
339
340
341
342
343
344
345
346
347
348
349
350
351
352
353
354
355
356
357
358
359
360
361
362
363
364
365
366
367
368
369
370
371
372
373
374
375
376
377
378
379
380
381
382
383
384
385
386
387
388
389
390
391
392
393
394
395
396
397
398
399
400
401
402
403
404
405
406
407
408
409
410
411
412
413
414
415
416
417
418
419
420
421
422
423
424
425
426
427
428
429
430
431
432
433
434
435
436
437
438
439
440
441
442
443
444
445
446
447
448
449
450
451
452
453
454
455
456
457
458
459
460
461
462
463
464
465
466
467
468
469
470
471
472
473
474
475
476
477
478
479
480
481
482
483
484
485
486
487
488
489
490
491
492
493
494
495
496
497
498
499
500
501
502
503
504
505
506
507
508
509
510
511
512
513
514
515
516
517
518
519
520
521
522
523
524
525
526
527
528
529
530
531
532
533
534
535
536
537
538
539
540
541
542
543
544
545
546
547
548
549
550
551
552
553
554
555
556
557
558
559
560
561
562
563
564
565
566
567
568
569
570
571
572
573
574
575
576
577
578
579
580
581
582
583
584
585
586
587
588
589
590
591
592
593
594
595
596
597
598
599
600
601
602
603
604
605
606
607
608
609
610
611
612
613
614
615
616
617
618
619
620
621
622
623
624
625
626
627
628
629
630
631
632
633
634
635
636
637
638
639
640
641
642
643
644
645
646
647
648
649
650
651
652
653
654
655
656
657
658
659
660
661
662
663
664
665
666
667
668
669
670
671
672
673
674
675
676
677
678
679
680
681
682
683
684
685
686
687
688
689
690
691
692
693
694
695
696
697
698
699
700
701
702
703
704
705
706
707
708
709
710
711
712
713
714
715
716
717
718
719
720
721
722
723
724
725
726
727
728
729
730
731
732
733
734
735
736
737
738
739
740
741
742
743
744
745
746
747
748
749
750
751
752
753
754
755
756
757
758
759
760
761
762
763
764
765
766
767
768
769
770
771
772
773
774
775
776
777
778
779
780
781
782
783
784
785
786
787
788
789
790
791
792
793
794
795
796
797
798
799
800
801
802
803
804
805
806
807
808
809
810
811
812
813
814
815
816
817
818
819
820
821
822
823
824
825
826
827
828
829
830
831
832
833
834
835
836
837
838
839
840
841
842
843
844
845
846
847
848
849
850
851
852
853
854
855
856
857
858
859
860
861
862
863
864
865
866
867
868
869
870
871
872
873
874
875
876
877
878
879
880
881
882
883
884
885
886
887
888
889
890
891
892
893
894
895
896
897
898
899
900
901
902
903
904
905
906
907
908
909
910
911
912
913
914
915
916
917
918
919
920
921
922
923
924
925
926
927
928
929
930
931
932
933
934
935
936
937
938
939
940
941
942
943
944
945
946
947
948
949
950
951
952
953
954
955
956
957
958
959
960
961
962
963
964
965
966
967
968
969
970
971
972
973
974
975
976
977
978
979
980
981
982
983
984
985
986
987
988
989
990
991
992
993
994
995
996
997
998
999
1000

Detailed Description of the Preferred Embodiments

Figure 1 is a block diagram of a typical cell 100 showing a single base station 102 in communication with a plurality of mobile terminals 104, 106, 108. These devices communicate within the cell 100 using the TDD mode of WCDMA data communication as will be described in detail herein. Delay diversity within the cell 100 is accomplished by transmitting derived versions of the same signal from different antennas with at least one distinct delay between any two antennas. The derived versions of the signal are obtained by scaling the amplitude and/or phase shifting. The simplest form of delay diversity would involve simply transmitting the same signal from different antennas with at least one distinct delay between any two antennas.

Figure 2 is a diagram of a TDD radio frame 200 that may be transmitted by the communication system of Figure 1. The radio frame 200 can be seen to have a duration of 10 msec. The radio frame is divided into 15 equal time slots 202-210. Each of these time slots 202-210 is further divided into 2560 chip times T_c . The diagram of Figure 3 illustrates the structure of a single TDD time slot. The time slot includes a first group of data symbols 320 having 1104 chips. This first group corresponds to 69 data symbols for an exemplary spreading factor of 16. The first group is followed by a midamble 322 having 16 symbols for the exemplary spreading factor of 16. These midamble symbols are a predetermined training sequence similar to pilot symbols of FDD systems discussed herein above. The midamble 322 symbols are cyclically time shifted for different users in the cell as previously discussed. A second group of data symbols 324 having another 1104 chips follows the midamble 322. Finally, the second group of data symbols 324 is followed by a guard period 326 of 96 chips.

Figure 4(a) is a block diagram 400 illustrating one implementation of delay diversity at the base station 102 shown in Figure 1 and having K channels 402-406 and N antennas 408-412 suitable for accommodating the TDD mode of WCDMA data communication according to one embodiment of the present invention. Transmit antenna delay diversity is generally implemented with fixed delays for each antenna 408-412 (usually a 1-chip delay per antenna). Delay diversity, as stated above, is a method of transmit antenna diversity in which the same signal is transmitted from multiple antennas,

with each antenna having a different time delay for the same signal. With continued reference to Figure 4(a), the signals associated with each channel 402-406 are delayed at baseband, so different delays 414-430 can be used for each channel 402-404.

Figure 4(b) is a block diagram 450 illustrating one implementation of delay diversity at the base station 102 shown in Figure 1 and having K channels 402-406 and N antennas 408-412 suitable for accommodating the TDD mode of WCDMA data communication according to one embodiment of the present invention. As stated above, delay diversity within the cell 100 is accomplished by transmitting derived versions of the same signal from different antennas 408-412 with at least one distinct delay 416, 418, 420, 424, 426, 428 between any two antennas. The derived versions of the signal are obtained by scaling the amplitude and/or phase shifting via multiplier/phase shifting elements 452 – 468. As also stated above, the simplest form of delay diversity would involve simply transmitting the same signal from different antennas with at least one distinct delay between any two antennas such as depicted in Figure 4(a).

As seen in Figure 5, the base station 102 must measure the delay profile 502, 504, 506 in the uplink in order to receive the uplink transmission from a mobile terminal 104, 106, 108. Therefore, no additional system complexity is required in measuring the positions of the paths in the uplink. Figure 5 is a block diagram 500 illustrating a method of choosing channel delays 414-430 associated with multiple antennas 408-412 according to one embodiment of the present invention. Since the cell 100 shown in Figure 1 is a time division duplex (TDD) system, the uplink and downlink profiles are identical. The delay 414-430 between the antennas 408-412 can be chosen so that the strongest signal paths between the base station 102 and the mobile terminals 104, 106, 108 do not overlap, thereby achieving full diversity. Thus, the present method is distinct from those presently known communication methods since it is adaptive in that the delay 414-430 is variable. The present method is also distinct from those presently known communication methods since it is associated with a TDD system. The delay 414-430 need not remain fixed for a particular channel 402, 404, 406, but can be altered accordingly in order to meet the needs of a changing path profile between the base station 102 and one or more mobile terminals 104, 106, 108. There is of course, less flexibility in choosing the delays between antennas for broadcast channels transmitted to all mobile terminals since the

same delay is used for all users. In this case, as stated herein before, the antenna delays 414-430 can be chosen so as to benefit the maximum number of users or the users at the edge of the cell.

Figure 6 is a diagram of a delay profile 600 at a mobile terminal 104, 106, 108 with a delay of 2 chips between two transmit antennas at the base station 102 shown in Figure 1 according to one embodiment of the present invention. When the channel impulse response, for example, has two paths separated by 1-chip and having relative powers of 0dB and -10dB respectively, the foregoing 2-chip delay will provide the signal delay profile depicted in Figure 6 at the mobile terminal. Since the paths from the two antennas do not overlap, there is a full 4-path diversity.

Figure 7 is a diagram of a delay profile 700 at a mobile terminal with a delay of only 1-chip between two transmit antennas at the base station shown in Figure 1 according to one embodiment of the present invention. When the above channel impulse response having two paths separated by 1-chip and further having relative powers of 0dB and -10dB respectively is transmitted with only a 1-chip delay, the signal delay profile 700 is seen by the mobile terminal. In this instance, the second path from antenna 1 overlaps with the first path from antenna 2, thereby resulting in less desirable 3-path diversity.

Space-time transmit diversity (STTD) is the open-loop transmit antenna diversity technique of choice for the frequency division duplex (FDD) mode of wideband CDMA data communication systems since it has more gain than delay diversity and does not introduce extra paths at the mobile terminal, thereby simplifying receiver complexity. STTD for TDD/WCDMA systems is discussed further in U.S. Patent Application entitled *Space Time Transmit Diversity For TDD/WCDMA Systems*, docket no. TI-28984, by Dabak et al., filed on February 25, 2000, and assigned to the assignee of the present invention, and is incorporated by reference herein. Further, joint detection using the zero-forcing block linear equalizer (ZF-BLE), which is a decorrelating detector, has been proposed for use at the mobile terminal in association with the time division duplex (TDD) mode of WCDMA data communication by A. Klein, G. Kaleh and P. Baier, *Zero Forcing and Minimum Mean-Square-Error Equalization for Multiuser Detection in Code-Division Multiple-Access Channels*, IEEE Trans. on Vehicular Technology, Vol.

45, No. 2, pp. 276-287 (May 1996). There are however, difficulties implementing the ZF-BLE when STTD is used since the ZF-BLE is a linear detector. These difficulties arise because conjugates of the data are transmitted on the second transmit antenna, and a linear detector can not efficiently handle conjugates of data. Further, there is a
 5 computational complexity increase of more than 100% which is not acceptable since computational power is limited at the mobile terminal 104, 106, 108.

The present delay diversity for the TDD mode of WCDMA data communication depicted in Figure 4 does not involve transmitting conjugates of data. Therefore, the receiver at the mobile terminal 104, 106, 108 can be implemented using the ZF-BLE with
 10 an increase in computational complexity of only about 2% to about 17% (depending on the number of users employing delay diversity) as shown in Table 1 below. The variability of computational complexity associated with the ZF-BLE shown in Table 1 is based upon 8-channels having delay diversity capabilities. The total impulse response of each channel 402, 404, 406, including the delay spread of the transmit antennas 408, 410,
 15 412 must be kept less than the length of the channel estimation window associated with the base station 102 such that signals received by a mobile terminal 104, 106, 108 via one antenna will not overlap signals received by the mobile terminal via a different antenna. Further, the delay spread of the antennas 408, 410, 412 associated with each channel 402, 404, 406 most preferably are limited to $M=16$ chips or less as shown by equation (1).

$$|Delay_{k,i} - Delay_{k,j}| \leq M \text{ chips}; i \in [1,2,\dots,N]; j \in [1,2,\dots,N] \quad (1)$$

Table 1

Total MIPS (non DD ZF-BLE)	Total MIPS (BCH DD ZF-BLE)	Total MIPS (all ch DD ZF-BLE)
74.49813333	76.05973333	86.9909333
MIPS increase (%)	2.0961599	
BCH DD ZF-BLE		
MIPS increase (%)	16.769279	
all ch DD ZF-BLE		

Figure 8 is a simulation diagram showing bit error rate (BER) performance with and without delay diversity on the downlink using the Vehicular B channel model and Figure 9 is a simulation diagram showing bit error rate (BER) performance with and without delay diversity on the downlink using the Outdoor-to-Indoor and Pedestrian channel model. There is a gain of 0.3dB at a BER of 0.10 with the Vehicular B channel and a gain of 2.4dB at a BER of 0.01 with the Outdoor-to-Indoor and Pedestrian channel.

This invention has been described in considerable detail in order to provide those skilled in the wireless communication art with the information need to apply the novel principles and to construct and use such specialized components as are required. In view of the foregoing descriptions, it should be apparent that the present invention represents a significant departure from the prior art in construction and operation. However, while particular embodiments of the present invention have been described herein in detail, it is to be understood that various alterations, modifications and substitutions can be made therein without departing in any way from the spirit and scope of the present invention, as defined in the claims which follow. For example, multiple variations in the order of symbol transmission can provide the delay diversity associated with the present invention. Further, the exemplary delay diversity of the present invention can be implemented in hardware or software, or a combination of both. In a preferred embodiment, the functions of a data communication system designed in conformance with the principals set forth herein are implemented as one or more integrated circuits, using a suitable processing technology, e.g., CMOS.

As another example, at least portions of the present invention may be implemented in computer programs, i.e. algorithms, executing on programmable computers each comprising a data processor, a data storage system, including both volatile and non-volatile memory and/or data storage devices, at least one input device, and at least one output device. Program code is applied to input data to perform the functions described herein and generate output information. The output information is applied to one or more output devices, in known fashion. Each such program may be implemented in a high level procedural or object oriented programming language to communicate with a computer system. However, the programs can be implemented in assembly or machine language, if desired. In any case, the language may be a compiled

